

# Control of Nonholonomic Electrically-Driven Tractor-Trailer Wheeled Robots based on Adaptive Partial Linearization

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## ABSTRACT

Wheeled Mobile Robots (WMRs) are simple, easy to move on hard and level terrain and can be controlled effectively. Due to these merits, many researchers have studied the challenges of WMRs. To improve the payload transportation capability of wheeled vehicles, one or several platform, named as trailer, may be towed to a tractor wheeled platform. In the current paper, for the first time, the motion control of such tractor trailer systems is addressed while the actuator dynamics is considered. Toward this goal, the system kinematics and dynamics will be derived and will be coupled to its actuators model. To control the considered nonholonomic system, the technique of input-output feedback linearization along with look-ahead point notion will be utilized. Besides, some of the imprecise parameters in the proposed model-based controller are identified in an on-line manner. The obtained computer simulation results support the soundness of the proposed controller.

## 1. Introduction

Robotic systems provide different services in various environments based on the mission requirements. The robotic systems may consist of dexterous manipulation and locomotion divisions. While the manipulation part provides the capability of the dexterous operation on the environment, the locomotion is responsible for the robot motion in the environment. Based on the environment, and the mission requirements, the robot locomotion may be different. The robots could be flying, move on the surface (ground) or travel underwater. The ground robots locomotion would mainly be wheeled, legged, tracked or some combination of the three aforementioned mechanisms.

Among various locomotion mechanisms of the robots, the wheels are simple, easy to control and fast for motion over hard even terrains. These advantages have encouraged the researchers to focus on solving the challenges of the wheeled robots, [1], [2]. While the single-platform wheeled robots are advantageous, their payload carrying capacity is limited. In order to improve the payload carrying capacity and preserving the system maneuverability, another wheeled platform(s) is (are) connected to the active platform. Such

wheeled robots are called tractor-trailer wheeled robots (TTWRs). The active platform which provide the overall system driving is called tractor while the other wheeled platforms which are passive are called trailers, see Fig. 1.

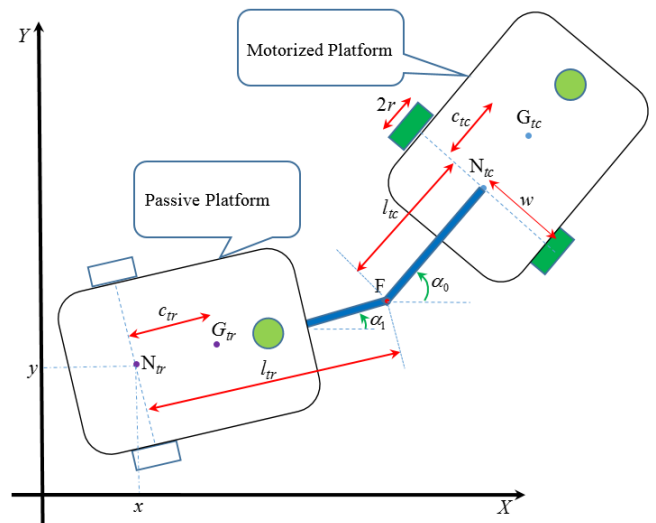


Figure 1. The considered tractor-trailer wheeled robot system

While such systems have been studied from the view point of motion planning [3-5], few works have been presented in the field of automatic control.

In the motion control field of the wheeled robots, there exist three types of problem including trajectory tracking, path following and posture stabilization [6]. The aim of trajectory tracking is that the desired trajectories associated with the coordinates of a reference point of the robot could be tracked. In [7], considering the trailer linear velocity and tractor angular velocity as control inputs, the trajectories of a reference point of the trailer have been controlled. To this end, the linear model predictive control technique has been utilized. In [8], first a Lyapunov-based kinematic controller has been designed for TTWRs. Then, a dynamic controller has been designed such that it can provide the required velocities of the kinematic controller using torques of the tractor wheels. In [9], the trajectory tracking of TTWRs has been realized using the notion of Modified Transpose Jacobian (MTJ) controller. One of the advantages of the use of MTJ is its simplicity and its independency from the robot dynamic model parameters. In [10], a TTWR is considered in which the trailer wheels are spherical. The considered under-actuated system is modeled and then is controlled using a physics based PID controller. The stability of the controller has been proven using the Lyapunov second method.

While few controllers have just been recently proposed to control TTWRs motion, most of them assume that the precise parameters of the robot are available. Besides, most of them ignore the actuators dynamics. Consequently, in the percent work, the considered TTWR is modeled and controlled while the actuator dynamics is taken into account. Besides, the values of some of the uncertain parameters of the robot and its actuators are identified. To control the robot, the input-output feedback linearization control technique is utilized adopting the look-ahead control strategy [11].

The rest of the present article is arranged as follows. In Section 2, the considered system is described and its kinematics and dynamics are derived. Then, in Section 3, the nonlinear developed controller as well as identification methods are introduced. In Section 4, the obtained simulation results are discussed. Some conclusions will be given in the last section.

## 2. system Description and its kinematics and dynamics

### 2.1. System Description

The considered system is composed of two wheeled platforms, as shown in Fig. 1. Each platform contains three wheels where two of them are conventional wheels and mounted at the rear of the platform. The third wheel is of type spherical and is utilized to enhance the equilibrium of the system. The wheeled platform whose rear wheels are active is named as the tractor while the other platform is called trailer. Since the two real wheels of the tractor are independently driven by electric DC motors, the tractor is of type differentially-driven platform and kinematically equivalent to unicycle type mobile robot, [12].

In Fig. 1, the various geometrical parameters of the robot is shown.  $N_{tc}$  and  $N_{tr}$  represent the midpoint of the rear axle of the tractor and the trailer, respectively. Also,  $x$  and  $y$  denote the Cartesian coordinates of point  $N_{tr}$  in inertial  $XY$  frame. Besides,  $\alpha_0$  and  $\alpha_1$  denote the heading angle of the tractor and the trailer with respect to  $X$  axis of inertial frame, respectively. Moreover,  $G_{tc}$  and  $G_{tr}$  denote the center of mass of the tractor and the trailer, respectively. In addition,  $w$  and  $r$  indicate the half of distance between center of wheels and the radius of wheels, respectively.

### 2.2. System Kinematics and Dynamics

To describe the system configuration, the following entities of vector  $\mu$  are selected

$$\mu = [x \quad y \quad \alpha_1 \quad \alpha_0]^T \quad (1)$$

It is recalled that the elements of vector  $\mu$  was explained in the former subsection. Herein, it is tried to examine the system kinematics and also derive the relation between the angular velocity of actuation wheels and the generalized velocities, i.e.  $\dot{\mu}$ . If the angular displacements of the tractor right and left active wheels are denoted by  $\delta_r$  and  $\delta_l$ , respectively, then the following relations can be written

$$\begin{aligned} v_{rw} &= r \dot{\delta}_r [\cos \alpha_0 \quad \sin \alpha_0]^T \\ v_{lw} &= r \dot{\delta}_l [\cos \alpha_0 \quad \sin \alpha_0]^T \end{aligned} \quad (2)$$

where  $v_{rw}$  and  $v_{lw}$  denote the linear velocity of the right and the left active wheels of the tractor, respectively. Notice that the above relations are correct provided that the wheels are subject to pure rolling condition. Also, as the linear velocity of the point  $N_{tc}$  relative to the tractor active wheels is perpendicular to the line connecting the aforementioned wheels centers, then it can be concluded that

$$\tan \alpha_0 = \dot{y}_{tc} / \dot{x}_{tc} \quad (3)$$

or

$$\dot{x}_{tc} \sin \alpha_0 - \dot{y}_{tc} \cos \alpha_0 = 0 \quad (4)$$

where  $x_{tc}$  and  $y_{tc}$  denote the Cartesian coordinates of the point  $N_{tc}$  in  $XY$  frame. Also, the following relations hold between coordinates of the points  $N_{tc}$  and  $N_{tr}$ .

$$\begin{aligned} x_{tc} &= x + l_{tr} c \alpha_1 + l_{tc} c \alpha_0 \\ y_{tc} &= y + l_{tr} s \alpha_1 + l_{tc} s \alpha_0 \end{aligned} \quad (5)$$

where  $c$  and  $s$  stand for functions  $\cos$  and  $\sin$ , respectively. By substituting (5) into (4), the following result is obtained.

$$\dot{x} \sin \alpha_0 - \dot{y} \cos \alpha_0 - \dot{\alpha}_1 l c (\alpha_0 - \alpha_1) = 0 \quad (6)$$

The recent relation is a non-integrable constraint and hence the TTWR is a nonholonomic mechanical system. By similar argument about the linear velocity of point  $N_{tr}$ , the following results is achieved.

$$\dot{x} \sin \alpha_1 - \dot{y} \cos \alpha_1 = 0 \quad (7)$$

The above relation is also nonholonomic which along with (6) can be written in the following compact matrix format.

$$\zeta(\boldsymbol{\mu})\dot{\boldsymbol{\mu}} = \mathbf{0} \quad (8)$$

in which

$$\zeta = \begin{bmatrix} \sin \alpha_0 & -\cos \alpha_0 & -l \cos(\alpha_0 - \alpha_1) & 0 \\ \sin \alpha_1 & -\cos \alpha_1 & 0 & 0 \end{bmatrix} \quad (9)$$

where  $l = l_{tr}$  if it is assumed that the on-axle hitching case is examined, i.e.  $l_{tc} = 0$ .

By writing the kinematic equation relating the linear velocity of the right and the left active wheels of tractor, then the following result is obtained.

$$\dot{\alpha}_0 = \frac{r}{2w}(\dot{\delta}_r - \dot{\delta}_l) \quad (10)$$

If  $v_{N_{tr}}$  is defined as the value (with sign) of the linear velocity of the point  $N_{tr}$ , then by velocity analysis of the points  $N_{tr}$  and  $F$ , the following result can be obtained.

$$v_{N_{tr}} = \frac{r}{2}(\dot{\delta}_r + \dot{\delta}_l) \cos(\alpha_0 - \alpha_1) \quad (11)$$

considering (10) and (11), the following imaginary kinematic input is defined as

$$\mathbf{u} = \begin{bmatrix} u_1 \\ u_2 \end{bmatrix} = \begin{bmatrix} v_{N_{tr}} \\ \dot{\alpha}_0 \end{bmatrix} = \mathbf{\Lambda} \begin{bmatrix} \dot{\delta}_r \\ \dot{\delta}_l \end{bmatrix} \quad (12)$$

$$\mathbf{\Lambda} = \begin{bmatrix} \frac{r}{2} \cos(\alpha_0 - \alpha_1) & \frac{r}{2} \cos(\alpha_0 - \alpha_1) \\ \frac{r}{2w} & -\frac{r}{2w} \end{bmatrix}$$

As seen the map between virtual kinematic input  $\mathbf{u}$  and active wheels' angular velocities is one-to-one. Hence, they are equivalent and can be utilized interchangeably. Since, the wheels of the trailer do not slip sideways, the following two relations can be written.

$$\begin{aligned} \dot{x} &= u_1 \cos \alpha_1 \\ \dot{y} &= u_1 \sin \alpha_1 \end{aligned} \quad (13)$$

Considering (10) the magnitude of the velocity of the point  $N_{tc}$  (with its sign) can be written in terms of the tractor wheels' velocities as

$$v_{N_{tc}} = \frac{r}{2}(\dot{\delta}_r + \dot{\delta}_l) \quad (14)$$

By recalling the point that in the present study,  $l_{tc} = 0$ , and writing the linear velocity relation between points  $N_{tc}$  and  $N_{tr}$ , and using (14), (13), and (12), the following result is concluded.

$$\dot{\alpha}_1 = \frac{1}{l} \tan(\alpha_0 - \alpha_1) u_1 \quad (15)$$

The relations (13), (15) and (10) using (12) can be written in the following compact form.

$$\dot{\boldsymbol{\mu}} = \mathbf{Q}(\boldsymbol{\mu}) \cdot \mathbf{u} \quad (16)$$

where

$$\mathbf{Q}^T = \begin{bmatrix} c \alpha_1 & s \alpha_1 & \frac{1}{l} \tan(\alpha_0 - \alpha_1) & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad (17)$$

As seen, (16) indicates that the first-order kinematics of the system is driftless, [12]. Also, one can conclude that the two independent columns of matrix  $\mathbf{Q}$  span the admissible generalized velocities  $\dot{\boldsymbol{\mu}}$ .

Now, the system dynamics model is extracted. In this regard, first the system Lagrangian is obtained. As the robot system has planar motion its Lagrangian  $L$  can be written as below if the wheels' mass is ignored.

$$L(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}}) = \frac{1}{2} m_{tc} v_{G_{tc}}^2 + \frac{1}{2} I_{tc} \dot{\alpha}_0^2 + \frac{1}{2} m_{tr} v_{G_{tr}}^2 + \frac{1}{2} I_{tr} \dot{\alpha}_1^2 \quad (18)$$

where  $m_{tc}$  and  $m_{tr}$  represent the tractor and the trailer masses, respectively. Also,  $I_{tc}$  and  $I_{tr}$  represent the tractor and the trailer mass moments of inertia about their associated center of mass, respectively. Besides,  $v_{G_{tc}}$  and  $v_{G_{tr}}$  denote the linear velocity value of center of masses of the tractor and the trailer, respectively. Note that if the right-hand side of the above equation is written in terms of  $\boldsymbol{\mu}$  and  $\dot{\boldsymbol{\mu}}$ , then by its substitution in the constrained Lagrange formulation, the following is resulted.

$$\mathbf{A}(\boldsymbol{\mu})\ddot{\boldsymbol{\mu}} + \mathbf{n}(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}}) \dot{\boldsymbol{\mu}} = \mathbf{E}(\boldsymbol{\mu})\mathbf{T} + \zeta^T(\boldsymbol{\mu})\boldsymbol{\lambda} \quad (19)$$

where  $\mathbf{A}$  represents the configuration dependent mass/inertia matrix and  $\mathbf{n}\dot{\boldsymbol{\mu}}$  is the vector containing the Coriolis and centripetal forces, [13]. Besides,  $\mathbf{E}$  is input matrix which transfers the effect of the wheels actuating torques  $\mathbf{T}$  to the generalized forces. Also,  $\boldsymbol{\lambda}$  denotes the  $2 \times 1$  vector of Lagrange multipliers. Now, by premultiplication of the above relation by  $\mathbf{Q}^T$ , and substituting of  $\dot{\boldsymbol{\mu}}$  by the right-hand side of (15), the following reduced dynamics model is achieved.

$$\bar{\mathbf{M}}\dot{\mathbf{u}} + \bar{\mathbf{C}}(\boldsymbol{\mu}, \mathbf{u})\mathbf{u} = \mathbf{Q}^T \mathbf{E}(\boldsymbol{\mu})\mathbf{T} \quad (20)$$

where

$$\begin{aligned} \bar{\mathbf{M}}(\boldsymbol{\mu}) &= \mathbf{Q}^T \mathbf{A}(\boldsymbol{\mu})\mathbf{Q} \\ \bar{\mathbf{C}}(\boldsymbol{\mu}, \mathbf{u}) &= \mathbf{Q}^T \mathbf{A}(\boldsymbol{\mu})\dot{\mathbf{Q}} + \mathbf{Q}^T \mathbf{n}(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}})\mathbf{Q} \end{aligned} \quad (21)$$

Note that (20) is obtained since the columns of the matrix  $\mathbf{Q}$  belong to the null-space of the matrix  $\zeta(\boldsymbol{\mu})$  and consequently,  $\zeta \mathbf{Q} = \mathbf{0}$ .

### 2.3. Actuator Model

The relation between the voltage source of armature circuit and its current can be described by the following first-order differential equation, [13].

$$l_A \dot{i}_A + R_A i_A = v_A - k_e \dot{\alpha}_m \quad (22)$$

where  $l_A$  and  $R_A$  denote, the inductance and resistance of the armature circuit, respectively. Besides,  $i_A$  and  $v_A$  denote the current and source voltage of the armature windings, respectively. Also,  $k_e$  indicate the back-electromotive force constant of the motor and  $\dot{\alpha}_m$  shows the angular velocity of the motor. In motors, the inductances are ignorable and the produced torque ( $\tau_m$ ) is related to the current flows in the armature windings as  $\tau_m = k_m i_A$ . Note that  $k_m$  is the motor torque constant. Hence, (20) is written as

$$\frac{R_A}{k_m} \tau_m = v_A - k_e \dot{\alpha}_m \quad (23)$$

If the motor is not direct drive and connected to the load via a gear reduction module then the following result is obtained

$$\frac{R_A}{K_m} \kappa \tau = v_A - \frac{k_e}{\kappa} \dot{\delta} \quad (24)$$

where  $\kappa > 1$  is the ratio of the gear reduction module. Moreover,  $\tau$  and  $\dot{\delta}$  denote the torque and angular velocity delivered to the load.

If it is assumed that the tractor motors and their associated gearboxes are similar, then by applying (24) to the right and the left motors of the tractor, the following result is achieved.

$$\mathbf{T} = \begin{bmatrix} \tau_r \\ \tau_l \end{bmatrix} = \gamma_1 \begin{bmatrix} v_{A_r} \\ v_{A_l} \end{bmatrix} - \gamma_2 \begin{bmatrix} \dot{\delta}_r \\ \dot{\delta}_l \end{bmatrix} \quad (25)$$

where

$$\gamma_1 = \frac{k_m}{\kappa R_A}, \quad \gamma_2 = \frac{k_e k_m}{\kappa^2 R_A} \quad (26)$$

Substituting (12) into (25), the following result is obtained.

$$\mathbf{T} = \begin{bmatrix} \tau_r \\ \tau_l \end{bmatrix} = \gamma_1 \begin{bmatrix} v_{A_r} \\ v_{A_l} \end{bmatrix} - \gamma_2 \mathbf{\Lambda}^{-1} \mathbf{u} \quad (27)$$

If the right hand side of the above relation is substituted for  $\mathbf{T}$  in (20), then the following relation is achieved.

$$\tilde{\mathbf{M}}(\boldsymbol{\mu}) \cdot \dot{\mathbf{u}}(t) + \tilde{\mathbf{C}}(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}}) \cdot \mathbf{u}(t) = \gamma_1 \tilde{\mathbf{B}}(\boldsymbol{\mu}) \cdot \mathbf{v}_a \quad (28)$$

where

$$\tilde{\mathbf{M}} = \bar{\mathbf{M}}$$

$$\mathbf{v}_a = \begin{bmatrix} v_{A_r} \\ v_{A_l} \end{bmatrix} \quad (29)$$

$$\tilde{\mathbf{B}}(\boldsymbol{\mu}) = \mathbf{Q}^T \mathbf{E}(\boldsymbol{\mu})$$

$$\tilde{\mathbf{C}} = \bar{\mathbf{C}}(\boldsymbol{\mu}, \mathbf{u}) + \gamma_2 \mathbf{Q}^T \mathbf{E}(\boldsymbol{\mu}) \mathbf{\Lambda}^{-1}$$

### 3. feedback linearization control and system Identification

In order to design a trajectory tracking controller for the system, the dynamical equation of the system should be considered. The  $\tilde{\mathbf{M}}$ ,  $\tilde{\mathbf{C}}$  and  $\tilde{\mathbf{B}}$  matrices, can be expressed as follows.

$$\begin{aligned} \tilde{\mathbf{M}}(\boldsymbol{\mu}) &= \begin{bmatrix} M_s + \frac{I_{tr}}{l^2} \tan^2(\alpha_0 - \alpha_1) & 0 \\ 0 & I_{tc} \end{bmatrix} \\ \tilde{\mathbf{C}}(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}}) &= \begin{bmatrix} c_{11} & -c_{12} \\ c_{12} & 2\gamma_2 \frac{w^2}{r^2} \end{bmatrix} \\ \tilde{\mathbf{B}}(\boldsymbol{\mu}) &= \frac{1}{r} \begin{bmatrix} \frac{1}{\cos(\alpha_0 - \alpha_1)} & \frac{1}{\cos(\alpha_0 - \alpha_1)} \\ w & -w \end{bmatrix} \end{aligned} \quad (30)$$

where

$$M_s = m_{tc} + m_{tr}$$

$$c_{12} = c_{tc} m_{tc} \frac{\dot{\alpha}_0}{\cos(\alpha_0 - \alpha_1)}$$

$$c_{11} = -\frac{I_{tr}}{l^2} \dot{\alpha}_1 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)} + \quad (31)$$

$$\frac{I_{tr}}{l^2} \dot{\alpha}_0 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)} + 2 \frac{\gamma_2}{r^2} \frac{1}{\cos^2(\alpha_0 - \alpha_1)}$$

Considering the kinematics and dynamics of the system, i.e. (28) and (16), the following state space representation will be obtained in control affine arrangement.

$$\dot{\boldsymbol{\eta}} = \begin{bmatrix} \mathbf{Q}\mathbf{u} \\ -\tilde{\mathbf{M}}^{-1} \tilde{\mathbf{C}}\mathbf{u} \end{bmatrix} + \begin{bmatrix} 0 \\ \gamma_1 \tilde{\mathbf{M}}^{-1} \cdot \tilde{\mathbf{B}} \end{bmatrix} \mathbf{v}_a \quad (32)$$

where  $\boldsymbol{\eta}^T = [\boldsymbol{\mu}^T \quad \mathbf{u}^T]$  is the state vector. This representation helps us to solve the trajectory tracking control problem via feedback linearization approach.

The above equation can be rewritten as follows.

$$\dot{\boldsymbol{\eta}} = h(\boldsymbol{\eta}) + b(\boldsymbol{\eta}, \boldsymbol{\theta}) + f(\boldsymbol{\eta}, \boldsymbol{\theta}) \mathbf{v}_a \quad (33)$$

where we have

$$h(\boldsymbol{\eta}) = \begin{bmatrix} \mathbf{Qu} \\ 0 \end{bmatrix}, b(\boldsymbol{\eta}, \boldsymbol{\theta}) = \begin{bmatrix} 0 \\ J(\boldsymbol{\eta}, \boldsymbol{\theta}) \end{bmatrix}, \quad (34)$$

$$f(\boldsymbol{\eta}, \boldsymbol{\theta}) = \gamma_1 \begin{bmatrix} 0 \\ F(\boldsymbol{\eta}, \boldsymbol{\theta}) \end{bmatrix}$$

The goal of control system is that using voltages applied to the wheels motors, i.e.  $\mathbf{v}_a$ , as control inputs, the trajectory tracked by point  $N_{tr}$ , i.e.  $[x(t) \ y(t)]$ , be as close as possible to the desired reference trajectory. Consequently,  $\mathbf{v}_a$  should be adjusted so as

$$\lim_{t \rightarrow \infty} (x_{des}(t) - x(t)) = 0, \quad \lim_{t \rightarrow \infty} (y_{des}(t) - y(t)) = 0 \quad (35)$$

The following output are selected as a desired trajectory to be tracked by the system based on look-ahead control approach, [11].

$$\mathbf{C} = \begin{bmatrix} x + D \cos(2\alpha_0 - \alpha_1) \\ y + D \sin(2\alpha_0 - \alpha_1) \end{bmatrix} \quad (36)$$

It is recalled that if  $D$  in (36), which is named as look-ahead distance, is considered to be zero, then the system will not be input-output linearizable, based on the next equations.

It should be noted that, it is assumed the system is controllable, the internal dynamics of the system is stable, and choosing an appropriate set of output equations, cause the system to be input-output linearizable.

The known approach to find an input-output equation is to frequently differentiate the outputs so that they are explicitly related to inputs. After differentiating (36), the following equations will be obtained.

$$\ddot{\mathbf{C}} = L_h^2 \mathbf{C}(\boldsymbol{\eta}) + L_b L_h \mathbf{C}(\boldsymbol{\eta}) + L_f L_h \mathbf{C}(\boldsymbol{\eta}) \cdot \mathbf{v}_a \quad (37)$$

where Lie algebra is utilized in the above equation. Now, the input  $\mathbf{v}_a$  can be chosen as

$$\mathbf{v}_a = (L_f L_h \mathbf{C}(\boldsymbol{\eta}))^{-1} [\ddot{\boldsymbol{\xi}} - L_h^2 \mathbf{C}(\boldsymbol{\eta}) - L_b L_h \mathbf{C}(\boldsymbol{\eta})] \quad (38)$$

The above non-linear feedback, linearizes and decouples the system in the following form.

$$\ddot{\mathbf{C}} = \boldsymbol{\xi} \quad (39)$$

where  $\boldsymbol{\xi}$  represents the virtual input vector. It can be chosen as

$$\boldsymbol{\xi} = \ddot{\mathbf{C}}_{des} + \mathbf{K}_1 \dot{\mathbf{e}} + \mathbf{K}_2 \mathbf{e} \quad (40)$$

where  $\mathbf{K}_1$  and  $\mathbf{K}_2$  are appropriate control gain matrices.

It is worth mentioning that the developed controller (38) is model-based and depends on the robot parameters. In the real world applications, the exact values of the parameters  $m_{tr}$ ,  $m_r$ ,  $r$ ,  $w$  and  $l$ , are available while the values of parameters related to position of the center of mass, moment of inertia and actuators' constants, such as  $c_{tc}$ ,  $I_{tc}$ ,  $I_r$ ,  $\gamma_1$  and  $\gamma_2$ , contain uncertainties. Therefore, the exact values of the mentioned parameters will be identified in this study. To this

end, the following parameters are considered as unknown values.

$$\boldsymbol{\theta} = [\gamma_1 \quad I_r / l^2 \quad I_{tc} \quad \gamma_2 \quad c_{tc} m_{tc}]^T \quad (41)$$

Applying this notation, Eq. (30) can be rewritten as follows.

$$\tilde{\mathbf{M}}(\boldsymbol{\mu}) = \begin{bmatrix} M_s + \theta_2 \tan^2(\alpha_0 - \alpha_1) & 0 \\ 0 & \theta_3 \end{bmatrix}$$

$$\tilde{\mathbf{C}}(\boldsymbol{\mu}, \dot{\boldsymbol{\mu}}) = \begin{bmatrix} c_{11} & -c_{12} \\ c_{12} & 2\theta_4 \frac{w^2}{r^2} \end{bmatrix} \quad (42)$$

$$\tilde{\mathbf{B}}(\boldsymbol{\mu}) = \frac{1}{r} \begin{bmatrix} 1 & 1 \\ \cos(\alpha_0 - \alpha_1) & \cos(\alpha_0 - \alpha_1) \\ w & -w \end{bmatrix}$$

where  $c_{12} = \theta_5 \frac{\dot{\alpha}_0}{\cos(\alpha_0 - \alpha_1)}$  and

$$c_{11} = -\theta_2 \dot{\alpha}_1 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)} + \theta_2 \dot{\alpha}_0 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)} + 2 \frac{\theta_4}{r^2} \frac{1}{\cos^2(\alpha_0 - \alpha_1)}$$

Now, the appropriate  $z_i$  variables are defined as,

$$z_1 = \tan^2(\alpha_0 - \alpha_1), \quad z_2 = \dot{\alpha}_1 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)},$$

$$z_3 = \dot{\alpha}_0 \frac{\sin(\alpha_0 - \alpha_1)}{\cos^3(\alpha_0 - \alpha_1)}, \quad z_4 = \frac{2}{r^2} \frac{1}{\cos^2(\alpha_0 - \alpha_1)},$$

$$z_5 = \frac{\dot{\alpha}_0}{\cos(\alpha_0 - \alpha_1)} \text{ and } z_6 = \frac{1}{\cos(\alpha_0 - \alpha_1)}.$$

Substituting above equations into (28) and applying additional mathematical manipulation and some simplifications will result to the following equations.

$$\begin{cases} \theta_2 Z_2 + \theta_4 Z_4 - \theta_5 Z_5 - \theta_1 Z_1 = -M_s \dot{u}_1 = y_s \\ \theta_3 Z_3 + \theta_5 Z_7 + \theta_4 Z_6 - \theta_1 Z_8 = 0 \end{cases} \quad (43)$$

where  $Z_i$  are appropriate regressors, and  $\theta_i$  are unknown parameters that should be identified *online* during system control. In order to identify these parameters, the Recursive least squares (RLS) estimation method are implemented [14]. In this method, the following equations is utilized to estimate the unknown parameters vector  $\boldsymbol{\theta}$ .

$$\left\{ \begin{aligned} \hat{\boldsymbol{\theta}}(m+1) &= \hat{\boldsymbol{\theta}}(m) + \\ &\frac{\mathbf{P}(m) \cdot \mathbf{Z}(m+1)}{1 + \mathbf{Z}^T(m+1) \cdot \mathbf{P}(m) \cdot \mathbf{Z}(m+1)} \boldsymbol{\varepsilon}(m+1) \\ \mathbf{P}(m+1) &= \mathbf{P}(m) - \\ &\frac{[\mathbf{P}(m) \cdot \mathbf{Z}(m+1)] [\mathbf{Z}^T(m+1) \cdot \mathbf{P}(m)]}{1 + \mathbf{Z}^T(m+1) \cdot \mathbf{P}(m) \cdot \mathbf{Z}(m+1)} \end{aligned} \right. \quad (44)$$

welchere  $\boldsymbol{\varepsilon}(m+1) = y_s(m+1) - \mathbf{Z}^T(m+1) \cdot \boldsymbol{\theta}(m)$ ,  $m$  is the preceding time index,  $\mathbf{P}$  is the covariance matrix,  $\mathbf{Z}$  is the regressors vector and  $\hat{\boldsymbol{\theta}}$  is the estimated unknown parameters vector. It should be noted that, all states of the system are available in real time from measurement.

It should be noted that because of identification process of some of the system parameters, the Eq. (38) should be utilized as follows in the control loop.

$$\mathbf{v}_a = \left( L_f L_h C(\boldsymbol{\eta}) \right)^{-1} \left[ \xi - L_h^2 C(\boldsymbol{\eta}) - L_f L_h C(\boldsymbol{\eta}) \right] \quad (45)$$

The above control input is the final control command to force the system to track the desired trajectory. Figure 2 shows the feedback linearizing control system for the trajectory tracking considering identification of the unknown parameters in the system.

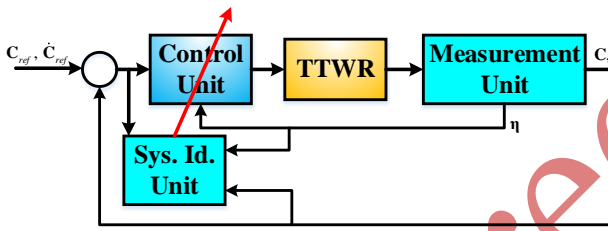


Figure 2. The control schematic of the TTWR system

#### 4. Simulation results

In this part, the simulation results of the implemented controller will be analyzed. The physical parameters of the system are reported in Table 1.

Table 1. The physical parameters of the system

Param*	Value	Unit	Param*	Value	Unit
$m_{ts}, m_{tr}$	0.9, 0.33	kg	$c_{ts}, c_{tr}$	0.29, 0	m
$I_{ts}, I_{tr}$	0.0035, 0.00078	kg.m <sup>2</sup>	$\gamma_1, \gamma_2$	0.2615, 0.2668	N.m/V, N.ms $\Omega^2$ /rad
$l$	0.17	m	$w$	0.0595	m
$r$	0.026	m	$\mathbf{K}_1, \mathbf{K}_2$	diag(1,1), diag(2,2)	--

\*: "Param" stands for "Parameter"

The desired reference trajectory is considered as a circular path as follows.

$$\begin{bmatrix} x_{des} \\ y_{des} \end{bmatrix} = \begin{bmatrix} 1.7 \cos(0.001t^2 + 0.1) \\ 1.7 \sin(0.001t^2 + 0.1) \end{bmatrix} \quad (46)$$

Notice that based on the above selection for the trajectories of point  $N_{tr}$ , the desired outputs are required to be calculated according to (36). In the next simulations the look-ahead distance  $D=0.17$ .

The initial condition of the robot is considered as  $\boldsymbol{\mu} = [-1.8597 \text{ m}, 0.1941 \text{ m}, -0.1 \text{ rad}, 0.5 \text{ rad}]$  as an arbitrary point in  $XY$  plane. Figure 3 shows the trajectory tracking performance of the trailer robot. Figure 4 and Figure 5 show the x and y components of the reference trajectory tracking errors, respectively. It is obvious that the desired trajectory was tracked accurately by the developed controller. Figure 6 and Figure 7 demonstrate the first and the second actuators' output, respectively. Considering Eq. (41), Figure 8 to Figure 12 show the parameters identification process by the adaptation algorithm. It is emphasized that in the simulations, the exact value of these parameters are assumed as  $\boldsymbol{\theta} = [0.2615, 0.90, 0.0792, 0.2668, 0.2610]^T$ . The initial guess of the  $\boldsymbol{\theta}$  in adaptation algorithm is considered as  $\hat{\boldsymbol{\theta}}(0) = [8, 10, 0.5, 0.4, 0.4]^T$ .

As can be observed in Figure 8 to Figure 12, the considered identification law has estimated all parameters rapidly in less than 1 sec. The final values of the estimated parameters by the developed identification method are as  $\hat{\boldsymbol{\theta}}_{identified} = [0.2626, 0.9062, 0.0802, 0.2679, 0.2621]^T$ .

Comparison of  $\hat{\boldsymbol{\theta}}_{identified}$  with the exact values of these parameters,  $\boldsymbol{\theta} = [0.2615, 0.90, 0.0792, 0.2668, 0.2610]^T$ , starting from  $\hat{\boldsymbol{\theta}}(0) = [8, 10, 0.5, 0.4, 0.4]^T$  prove the appropriate performance of the employed identification approach.

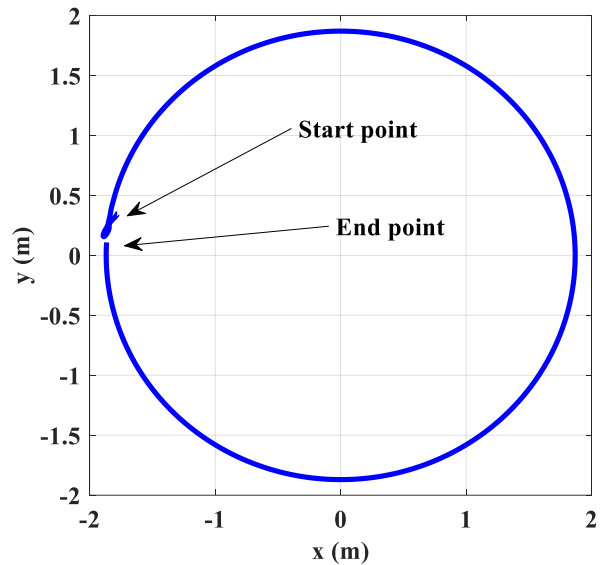


Figure 3. The path tracked by the robot trailer

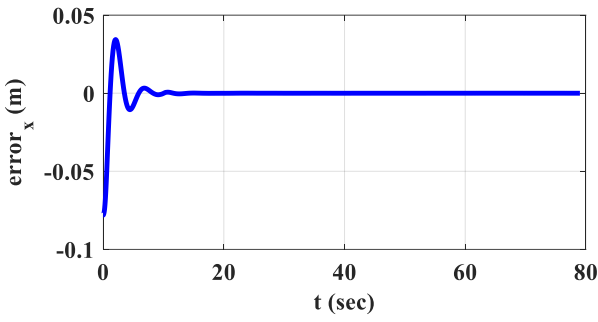


Figure 4. The x component of the reference trajectory tracking error

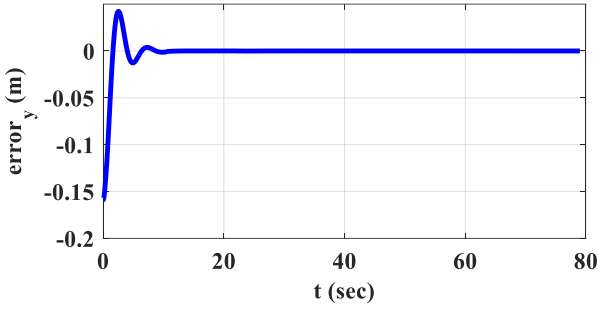


Figure 5. The y component of the reference trajectory tracking error

In order to simulate the real world condition for the TTWR system, the disturbance attenuation performance of the system is analyzed in the simulation tests. To this end, 50% of the control input was added to the system original input to test the disturbance attenuation performance of the developed controller. Figure 13 and Figure 14 show x and y components of the reference trajectory tracking errors considering disturbed input, respectively. Two sets of other simulations considering step-shape and sin-shape disturbances were conducted in the implemented tests. Figure 15 and Figure 16 show x and y components of the reference trajectory tracking errors considering step-shape disturbed input, respectively. The amplitude of this disturbance was 1 mV. Figure 17 and Figure 18 show x and y components of the reference trajectory tracking errors considering sine-shape disturbed input, respectively. The amplitude of this disturbance was 0.5 mV. Moreover, 10% of uncertainty in the masses of the first and the second platforms was added to the system. Figure 19 and Figure 20 show x and y components of the reference trajectory tracking errors considering this type of uncertainty.

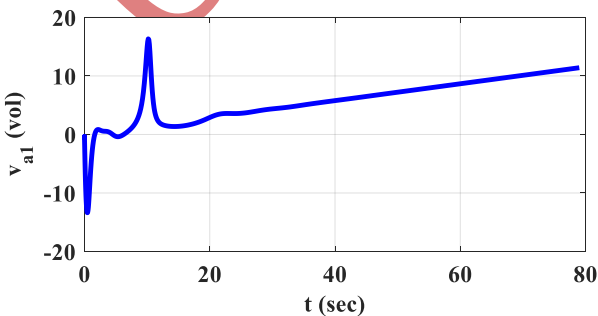


Figure 6. The time history of the first actuator output

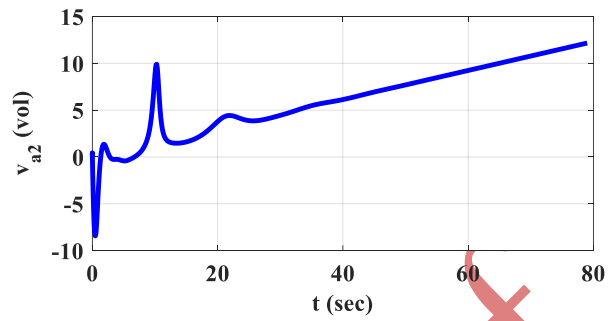


Figure 7. The time history of the second actuator output

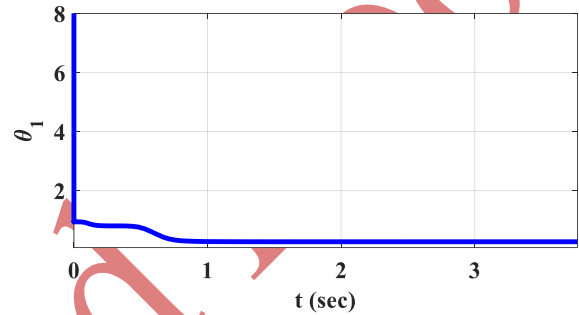


Figure 8. The  $\theta_1$  convergence time history

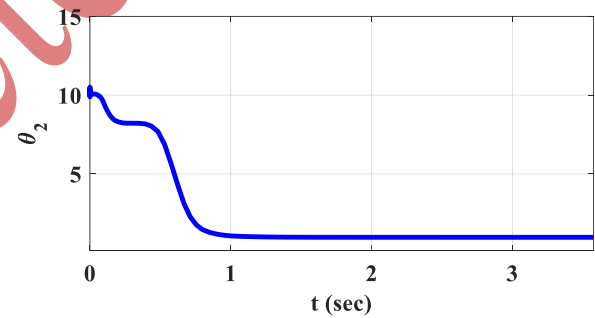


Figure 9. The  $\theta_2$  convergence time history

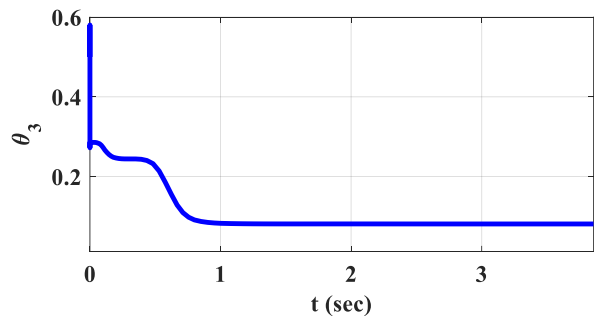


Figure 10. The  $\theta_3$  convergence time history

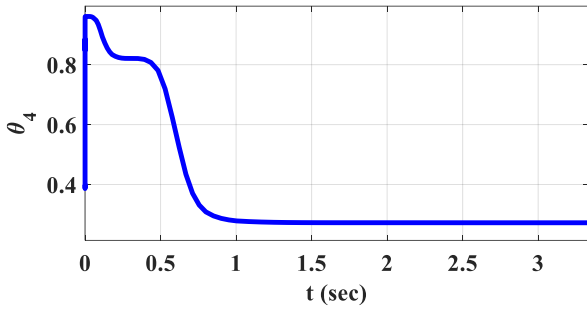


Figure 11. The  $\theta_4$  convergence time history

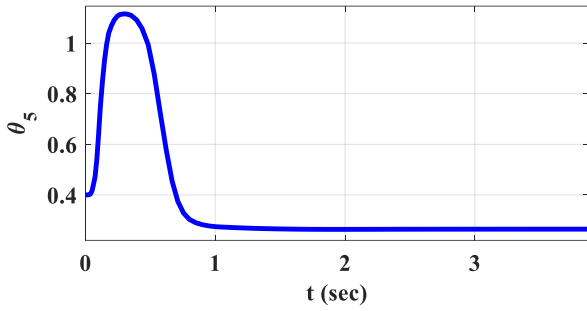


Figure 12. The  $\theta_5$  convergence time history

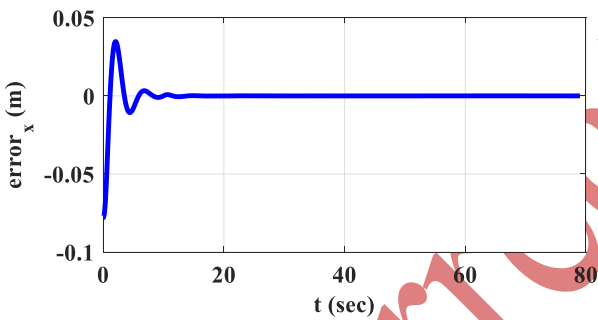


Figure 13. The x component of the reference trajectory tracking error considering disturbed input

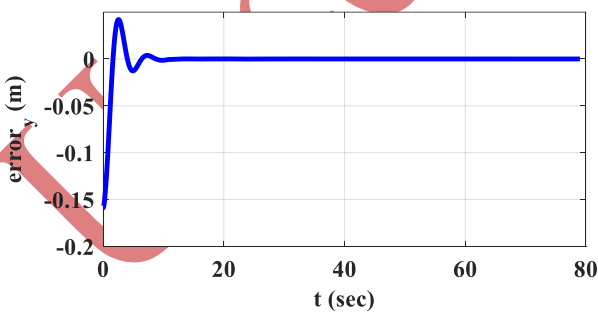


Figure 14. The y component of the reference trajectory tracking error considering disturbed input

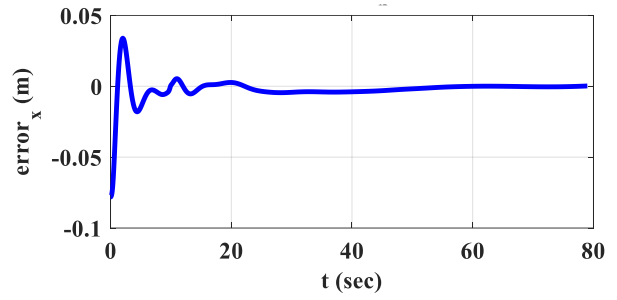


Figure 15. The x component of the reference trajectory tracking error considering step-shape disturbed input

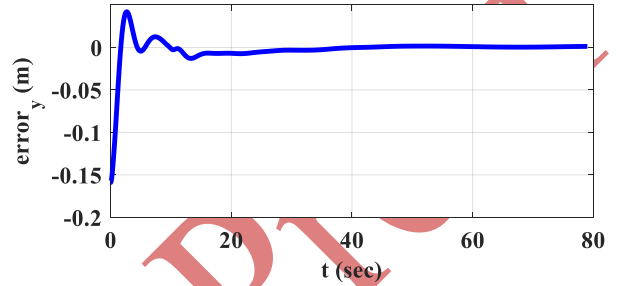


Figure 16. The x component of the reference trajectory tracking error considering step-shape disturbed input

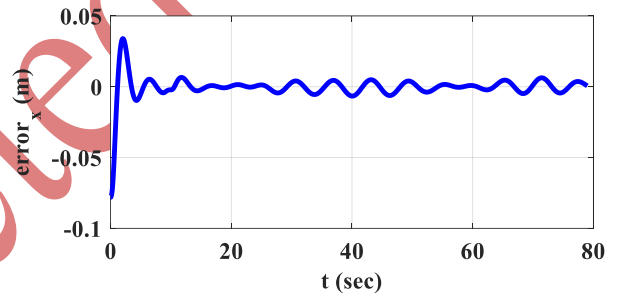


Figure 17. The x component of the reference trajectory tracking error considering sin-shape disturbed input

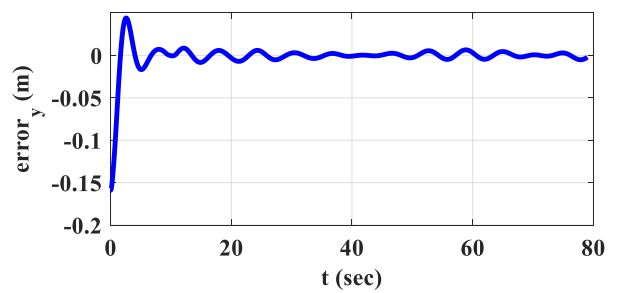


Figure 18. The y component of the reference trajectory tracking error considering sin-shape disturbed input



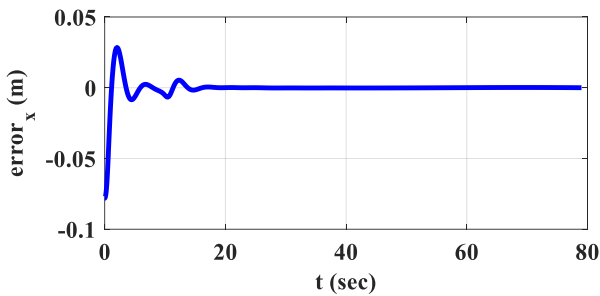


Figure 19. The x component of the reference trajectory tracking error considering uncertainty

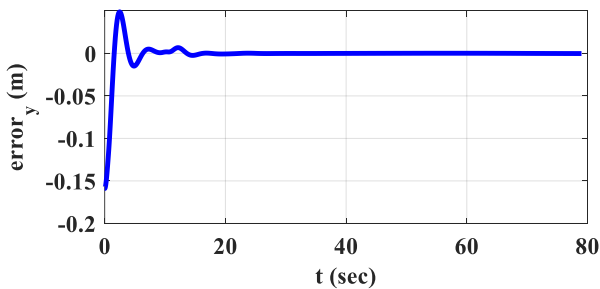


Figure 20. The y component of the reference trajectory tracking error considering uncertainty

## 5. Conclusions

In the present research, for the first time an adaptive partial feedback linearization approach was presented to trajectory tracking control of TTWRs. To this end, first an integrated model of the system kinematics, dynamics and actuator model was derived. Then, considering appropriate outputs using look-ahead method, the input-output feedback linearizing technique was employed. Since the exact value of some robot parameters may not be available in the controller, using RLS identification procedure, they were obtained. The obtained computer simulation results revealed the performance of the developed controller in terms of its accuracy and disturbance attenuation capability.

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